Demonstration of Linear Digital Filters (FIR)

A. Linear digital filters may be conceived of as vectors of weights that are to be multiplied by the digitally sampled values from a waveform. The filters given below are both 11 point digital filters with a half-amplitude frequency cutoff of approximately 17.5 Hz for data sampled at 200 Hz.

LOW PASS			HIGH PAS	S	
FILTER	COEFFICIENT	LAG	FILTER	COEFFICIENT	LAG
	0.0166	5	ĺ	-0.0166	5
	0.0402	4		-0.0402	4
	0.0799	3		-0.0799	3
	0.1231	2		-0.1231	2
	0.1561	1		-0.1561	1
	0.1684	0		0.8316	0
	0.1561	-1		-0.1561	-1
	0.1231	-2		-0.1231	-2
	0.0799	-3		-0.0799	-3
	0.0402	-4	İ	-0.0402	-4
	0.0166	-5		-0.0166	-5
			•		

So, what does all that fancy jargon mean?

1. The fact that these are 11 point filters indicates that 11 sample points are used in the determination of the new filtered value of any one sample point. Specifically, the sixth sample point is a weighted sum of the first 11 samples. In general, for an 11 point filter, any filtered sample point is the weighted sum of the preceding 5 sample values, the unfiltered sample value itself, and the subsequent 5 sample values. The lag refers to which sample point, relative to the center sample point, is to be weighted by the particular coefficient.

The <u>non-recursive</u> filter uses raw sample values in the calculations; <u>recursive</u> filters use the already filtered values of preceding samples in the calculations. Non-recursive filters are more straightforward and more commonly used.

2. The term <u>linear</u> denotes that the filter involves the computation of <u>weighted sums</u> of the digital sample values. Other filtering algorithms can be devised, but are not often applied to psychophysiological signals.

3. Digital filters, unlike analog filters, have characteristics that are sampling-rate dependent. These same filters would have a different cutoff frequency for data sampled at different sampling rates. Once you know the characteristics of a digital filter at a given frequency, it is a simple matter to convert the filter to another sampling rate as follows:

17.5/200 = x/1000 ; x = 87.5 @ 1000 Hz Sampling rate 17.5/200 = x/20 ; x = 1.75 @ 20 Hz Sampling rate B. So, you may ask, how do I work these things? First, consider a much simpler filter with the coefficients [.25 .50 .25]. This is a three point low pass filter. Consider the digitally sampled wave:

[0 2 6 4 10 4 6 2 0 -2 -6 -4 -10 -4 -6 -2 0] You should plot out these sample values as a function of time to examine the waveform.

1. The filter is applied as follows: the three filter coefficients are multiplied by the first three sample points and then summed to yield the middle sample point. So, .25*0 + .5*2 + .25*6 = 2.5; 2.5 is thus the new value for the second sample point. The filter "window" is then "shifted" or "slid" to the next 3 sample points and the procedure is repeated. Applying the filter to the entire sampled waveform yields:

[-- 2.5 4.5 6.0 7.0 6.0 4.5 2.5 0 -2.5 -4.5 -6.0 -7.0 -6.0 -4.5 -2.5 --]

Now, **you should plot out these filtered sample values on the same graph** you just constructed to examine the effects of filtering. Notice how the waveform has been smoothed.

2. Notice also that there exist no filtered values for the first and last samples. For most linear digital filters, some sample loss occurs. In general, the number of samples lost is P - 1, where P is the number of points in the filter; (P-1)/2 samples are lost at the beginning of the epoch and (P-1)/2 samples are lost at the end of the epoch.

If you like metaphors, this is the digital filter's "capacitor" which needs to charge. More precisely, the filter needs to take into account the trend or changes in the waveform; the discontinuities at the beginning and end of the epoch make such a trend analysis impossible. The most straightforward solution is therefore to refrain from filtering the earliest and latest samples in each epoch since the information required is unavailable.

3. You should apply this simple high-pass filter [-.25 .50 -.25] to the original waveform to see what happens. Plot the results.

C. The more elaborate 11 point filters are applied in the same manner. This will be demonstrated using basic matrix algebra. If the original waveform is represented by the matrix:

 $\mathbf{M} = [t_0, t_1, t_2, t_3, \dots, t_{n-1}]_{1 \times n}$

then the filtered matrix (where f_0 is actually at time point ((P-1)/2)+1):

 $\mathbf{F} = [f_0, f_1, f_2, f_3, \dots, f_{m-1}]_{1 \times m}$ (where m = n - (P-1))

is obtained by vector multiplication as follows:

1. First a matrix of shifted sample values is created:

```
S = [t_0, t_1, t_2, t_3, \dots, t_{10}

t_1, t_2, t_3, t_4, \dots, t_{11}

t_2, t_3, t_4, t_5, \dots, t_{12}

\dots

t_{n-11}, t_{n-10}, \dots, t_{n-1}]_{mxP}
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2. This matrix is then post-multiplied by the filter coefficient matrix C_{Px1} (C is either of the 11 point filter vectors listed at the top of the handout) to yield the filtered matrix of sample values:

$$\mathbf{F'}_{mx1} = \mathbf{S}_{mxP} * \mathbf{C}_{Px1}$$

3. The following pages of the handout contain plots of data filtered by the above procedure. First, the raw data are presented. Three waveforms were generated for a 20 Hz sampling rate (what's the Nyquist frequency?). The .5 Hz and 2 Hz waveforms were simply added together in phase to produce the more complex waveform which is to be filtered. The last two columns represent the high-pass and low-pass filtered versions of this more complex waveform. At a 20 Hz sampling rate, the 1/2-amplitude frequency cutoff of these filters is 1.75 Hz. Because the roll-off characteristics of these filters are rather broad, the original .5 & 2 Hz waveforms were not reproduced with fidelity. Nonetheless, you can see that fairly close approximations are produced. Plots of the original three waveforms and of the filtered versions follow.

4. In general, two features are important in determining a filter's length: roll off and 1/2 amplitude frequency. The sharper the roll-off desired, the longer the filter must be. The lower the 1/2 amplitude frequency, again the longer the filter. This last relationship should be intuitive: longer frequencies require more sample points to capture 1 wave period.

The last thing you should remember, now that you have the hang of digital filtering: digital filters are not always intuitive!

Generated	waveforms:	1	Ηz	&	4	Ηz	Sine	e waves	and	linear	combination
	Filtered	versions		S	of	the	he linear		ination		

				Filte	Filtered						
TIME (msec)	.5 Hz	2 Hz	SUM	High Pass	Low Pass						
0	0.0	0.0	0.0								
50	.156	.588	.744								
100	.309	.951	1.260								
150	.454	.951	1.405								
200	.588	.588	1.176								
250	.707	.000	.707	.04103	.66596						
300	.809	588	.221	33555	.55655						
350	.891	951	060	56681	.50681						
400	.951	951	.000	56335	.56335						
450	.988	588	.400	32477	.72477						
500	1.000	000	1.000	.05826	.94173						
550	.988	.588	1.575	.43948	1.1355						
600	.951	.951	1.902	.67404	1.2280						
650	.891	.951	1.842	.67058	1.1714						
700	.809	.588	1.397	.42970	.96730						
750	.707	.000	.707	.04103	.66596						
800	.588	588	.000	34822	.34822						
850	.454	951	497	59229	.09529						
900	.309	951	642	60075	04125						
950	.156	588	431	37297	05803						
1000	.000	000	000	00000	.00000						
1050	156	.588	.431	.37297	.05803						
1100	309	.951	.642	.60075	.04125						
1150	454	.951	.497	.59229	09529						
1200	588	.588	.000	.34822	34822						
1250	707	.000	707	04103	66596						
1300	809	588	-1.397	42970	96730						
1350	891	951	-1.842	67058	-1.1714						
1400	951	951	-1.902	67404	-1.2280						
1450	988	588	-1.575	43948	-1.1355						
1500	-1.000	000	-1.000	05826	94173						
1550	988	.588	400	.32477	72477						
1600	951	.951	000	.56335	56335						
1650	891	.951	.060	.56681	50681						
1700	809	.588	221	.33555	55655						
1750	707	.000	707								
1800	588	588	-1.176								
1850	454	951	-1.405								
1900	309	951	-1.260								
1950	156	588	744								
2000	000	000	000								

P.S.	:	Answer	to	High	Pass	Filt	ter	Exce	erc	ise	2							
		[5	5 1.5	-2 3	-2 2	1.5	5	0	.5	-1.5	2	-3	2	-1.5	.5]	